

Three phase induction motor using direct torque control by Matlab Simulink

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ABSTRACT

Induction motors are widely used in industries due to their rugged structure, high maintainability and economy than DC motors. There has been constant development in the induction motor drive system and their implementation in industrial applications. Direct Torque Control (DTC) is one of the control strategies of torque control or speed control in induction machine. It is considered as an alternative to the field oriented control or vector control techniques. These two control strategies are different on the operation principle but their objectives are the same. In this method the stator voltage vectors is selected according to the differences between the reference and actual torque and stator flux linkage. Here only the stator current and voltages is needed to calculate the actual torque and the flux linkages. The status of the torque and flux linkages applied the optimal switching logic. This optimal switching logic is used for the switching of the three-phase inverter. The inverter output gives the controlled voltage vectors that is applied to the induction motor. By properly adjusting the inverter switching the speed of the motor is controlled. The performance of an induction motor under the classical Direct Torque Control method and improved scheme have been studied and confirmed by simulation using MATLAB.

KEYWORDS:- Pulse-width modulation, direct field oriented control, digital signal processor, direct signal processor, direct torque control, application specific Integrated Circuit (ASIC).

1.INTRODUCTION

Industrial loads require operation at wide range of speeds. Such loads are generally termed as variable speed drives. These drives demand precise adjustment of speed in a steeples manner over the complete speed range required. The loads may be constant torque or a function of speed. These loads are driven by hydraulic, pneumatic or electric motors. An industrial drive has some special features when driven by electric motors. Induction machines have provided the most common form of electromechanical drive for industrial, commercial and domestic applications that is operate at essentially. In Induction motor 3-phase power supply provides a rotating magnetic field in an induction motor. The basic difference between an induction motor and a synchronous AC motor is that in the latter a dc current is supplied onto the rotor. Then magnetic field of stator which, through magnetic interaction, links to the rotating magnetic field in the stator which in turn causes the rotor to turn. It is called synchronous because at steady state the speed of the rotor is the same as the speed of the rotating magnetic field in the stator. By way of contrast, the induction motor does not have any direct supply onto the rotor; instead, a secondary current is induced in the rotor. To achieve this, stator windings are arranged around the rotor so that when energized with a poly-phase supply they create a rotating magnetic field pattern which sweeps past the rotor. This changing magnetic field pattern is induce currents in the rotor conductors. These currents interact with the rotating magnetic field created by the stator and the rotor produce torque.

However, for rotor current to be induced, the speed of the physical rotor and the speed of the rotating magnetic field in the stator must be different, or else the magnetic field will not be moving relative to the rotor conductors and no currents will be induced. Direct Control (DTC) are two types of drives employed for high performance applications. Direct Torque Control was introduced in Japan by Takahashi (1984) and Depenbrock (1985). Vector controlled induction motors are employed in high performance drives having precise speed control and good static as well as dynamic response. Direct Torque Controlled drives have increasingly become the best alternative to Field-Oriented Control methods. Modern control methods use state space techniques. The method of stabilizing the drives and improvement in their transient responses have been realized by modern power electronic devices. The block diagram of Speed Control scheme for an induction motor is as shown in Fig. 1.

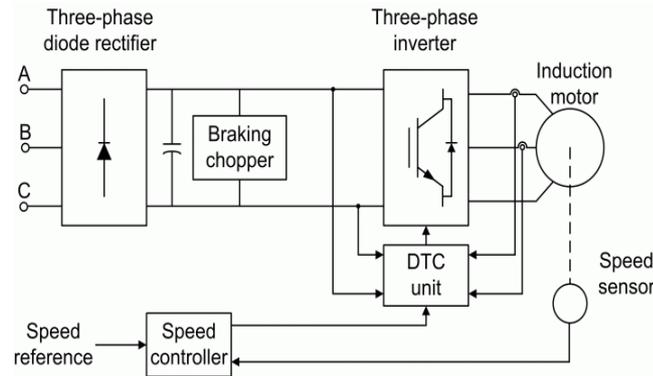


Fig. 1. Block Diagram of the Speed Control scheme

2.INDUCTION MOTOR

An induction motor is an AC electric motor in which the electric current in the rotor needed to produce torque is obtained by electromagnetic induction from the magnetic field of the stator winding. The objective of DTC is to control the induction motor. The per-phase equivalent circuit of an induction motor is valid only in steady-state condition. In an adjustable speed drive like the DTC drive, the machine normally constitutes an element within a feedback loop and hence its transient behavior has to be taken into consideration. The induction motor can be considered to be a transformer with short circuited and moving secondary. The coupling coefficients between the stator and rotor phases change continuously in the course of rotation of rotor. Hence the machine model can be described by differential equations with time-varying mutual inductances.

3.DIRECT TORQUE CONTROL

Direct torque control has its roots in field-oriented control and direct self-control. Field-oriented control uses spatial vector theory to optimally control magnetic field orientation. It has been successfully applied to the design of flux vector controls and is well documented. Direct self-control theory is less well known. The fundamental premise of direct self-control is as follows. Given a specific dc-link voltage and a specific stator flux level, a unique frequency of inverter operation is established. This is true because the time required by the time integral of the voltage to integrate up to the field flux level is unique and represents the half-period time of the frequency of operation. Since the operational frequency is established without a frequency reference, this operational mode is referred to as direct self-control. Output frequency is, thus, not requested, but rather, is self-controlled via the actual frequencies present. Once sensed, whether the frequency increases or decreases depends on what the torque reference from the speed regulator requests. Differential changes to operational frequency are determined by the torque request. Direct torque control combines field-oriented control theory, direct self-control theory, and recent advances in digital signal processor. DSP and application specific Integrated Circuit (ASIC) Technology to achieve a practical sensor less variable frequency drives.

4.VOLTAGE SOURCE INVERTER

Voltage source inverters (VSI) are commonly used to transfer real power from a DC power source to an AC load, such as an AC motor. Usually, the DC source voltage is nearly constant and the amplitude of the AC output voltage is controlled by adjusting the PWM ratio of the VSI. For a six-pulse VSI, according to its switch positions (S1 to S6), there are six non-zero active voltage space vectors (V1, V2, V3, V4, V5 and V6) and two zero voltage space vectors (V7 and V8) as shown in fig 2 One switch per leg of the VSI conduct at any time, i.e., if S1 is ON then S4 is OFF; 1 represents the ON state of a upper switch of a leg and 0 represents the ON state of the lower switch of the same leg. The stator flux linkage vector will move fast if non-zero switching vectors are applied and for a zero switching vector it will almost stop.

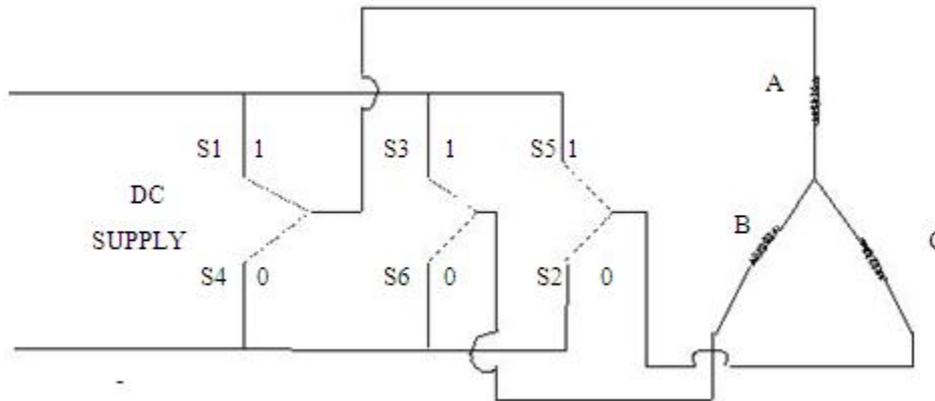


Fig.2.VSI Switching Position

5.LITERATURE REVIEW

DIRECT TORQUE CONTROL

Direct torque control (DTC) method is used in variable frequency drives to control the torque (and thus finally the speed) of three-phase AC electric motors. This involves calculating an estimate of the motor's magnetic flux and torque based on the measured voltage and current of the motor. Torque is estimated as a cross product of estimated stator flux linkage vector and measured motor current vector. The estimated flux magnitude and torque are then compared with their reference values. If either the estimated flux or torque deviates too far from the reference tolerance, the transistors of the variable frequency drive are turned off and on in such a way that the flux and torque errors will return in their tolerant bands as fast as possible. Thus direct torque control is one form of the hysteresis or bang-bang control. The DTC switches the inverter according to the load needs. Due to elimination of the fixed switching pattern (characteristic of the vector and the scalar control), the DTC response is extremely fast during the instant load changes. Corresponding to all possible combinations of switching states, active (non-zero) switching-voltage space vectors are shown in fig 3. By applying suitable space vector, changes in flux and torque demand can be met. If stator flux lies in sector and if a reduced stator flux linkage space vector modulus is required, the modulus is controlled by applying switching voltage vector which are directed towards the center.

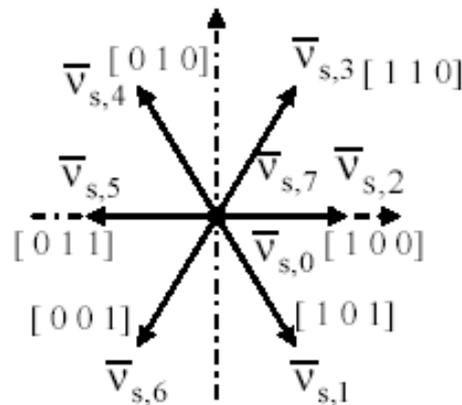


Fig.3 Voltage Switching Vectors

Torque is controlled by varying the angle between the stator flux vector and the rotor flux vector. This method is feasible because the rotor time constant is much larger than the stator time constant in reality, there are only six active voltage vectors and two zero-voltage vectors that a voltage-source inverter can produce. The analysis performed by the optimal switching logic is based on the mathematical spatial vector relationships of stator flux, rotor flux, stator current, and stator voltage.

Thus, rotor flux is relatively stable and changes quite slowly, compared to stator flux. When an increase in torque is required, the optimal switching logic selects a stator voltage vector that develops a tangential pull on the stator flux vector, tending to rotate it counterclockwise with respect to the rotor flux vector. The enlarged angle created effectively increases the torque produced. When a decrease in torque is required, the optimal switching logic selects a zero-voltage vector, which allows both stator flux and produced torque to decay naturally. If stator flux decays below its normal

lower limit the flux status output will again request an increase in stator flux. If the torque status output is still low, a new stator voltage vector is selected that tends to increase stator flux while simultaneously reducing the angle between the stator and rotor flux vectors.

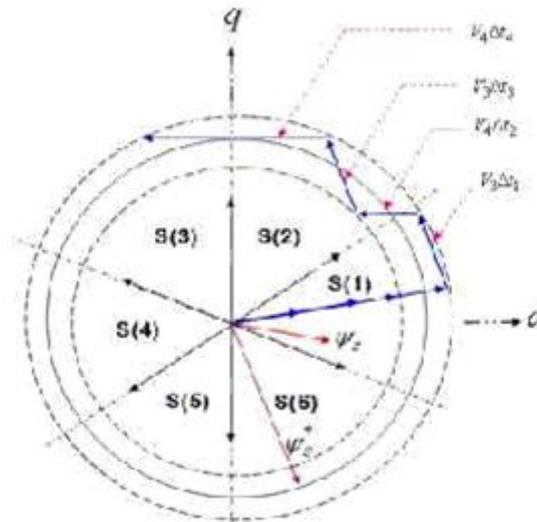


Fig.4: Trajectory of stator flux vector

Note that the combination of the hysteresis control block (torque and flux comparators) and the ASIC control block (Optimal switching logic) eliminate the need for a traditional PWM modulator. This provides two benefits. First, small signal delays associated with the modulator are eliminated and second, the discrete constant carrier frequencies used by the modulator are no longer present.

DIRECT TORQUE CONTROL DRIVE AND MODELING OF INDUCTION MOTOR

Torque Reference Controller-Within the Torque Reference Controller, the speed control output is limited by the torque limits and DC bus voltage. It also includes speed control for cases when an external torque signal is used. The internal torque reference from this block is fed to the torque comparator. The model and control blocks used to arrive at the firing signals for the VSI are described below:

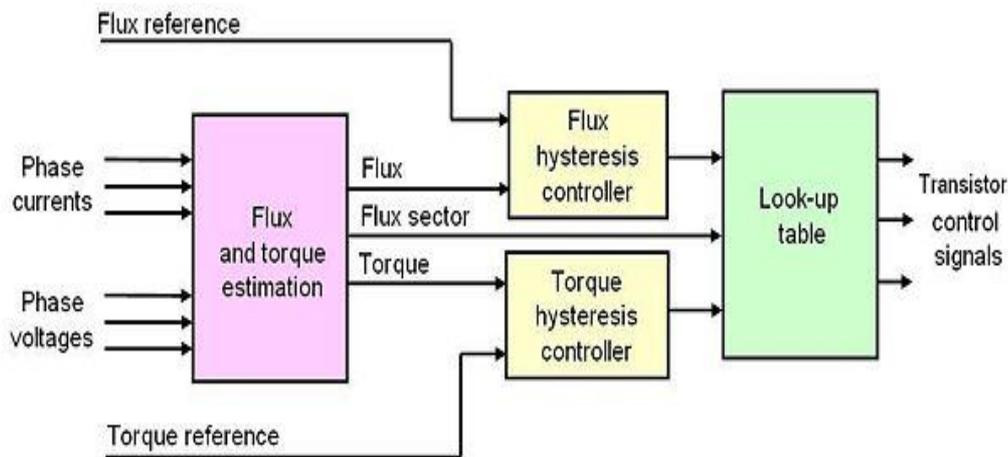


Fig.5. Direct Torque Control Drive.

Speed Controller- the Speed Controller block consists both of a PID controller and an acceleration compensator. The external speed reference signal is compared to the actual speed produced in the motor model. The error signal is then fed to both the PID controller and the acceleration compensator. The output is the sum of outputs from both of them. Flux Reference Controller- An absolute value of stator flux can be given from the flux reference controller to the flux comparator block. The ability to control and modify this absolute value provides an easy way to realize many inverter functions such as flux optimization and flux braking.

MODELING

Induction motor can be represented by the following equations in “dq0” format, in arbitrary reference frame, which is rotating at an angular speed in the direction of rotation of the rotor:

For electromagnetic torque calculation following equations is used:

$$T_{em} = (3/2)(p/2)(Lm / Lr) (dr Iqs - qr Ids) \tag{1}$$

$$T_{em} - T_{load} = J (d\omega_r / dt) \tag{2}$$

For motoring operation load torque is positive and for generating mode of operation load torque is negative.

MODELING OF FUNCTIONAL BLOCKS

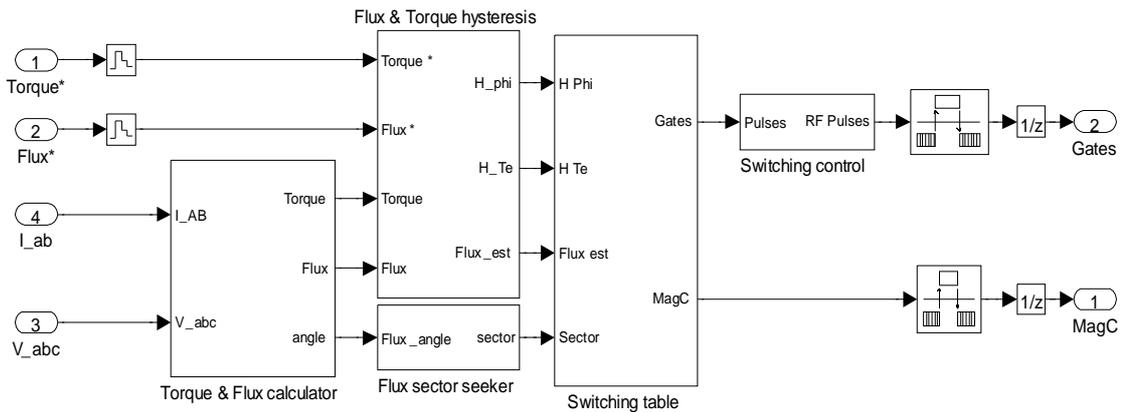


Fig. 6: Simulink diagram of DTC Model

TORQUE CALCULATION

Torque calculation is done using following equation

SPEED CALCULATION

Speed calculation is done using open loop estimator governed by following equation:

$$\omega_r_est = [\Psi_{dr} (d\Psi_{qr} / dt) - \Psi_{qr} (d\Psi_{dr} / dt)] [1/(\Psi_{dr}^2 + \Psi_{qr}^2)] - [(Lm / Tr)(\Psi_{dr} I_{qs} - \Psi_{qr} I_{ds})] [1/(\Psi_{dr}^2 + \Psi_{qr}^2)]$$

HYSTERESIS CONTROLLER

The speed regulator which generates the reference torque signal is compared against estimated torque and torque error is calculated in hysteresis comparator block. Similarly, flux error signal is also calculated. This is done using two level flux comparator and three level torque comparator using following consideration:

$$d\Psi=0 \text{ if } |\Psi_s| \geq |\Psi_s \text{ ref}| + |\Delta\Psi_s|$$

and

$$dt_e = 1 \text{ if } |te| \leq |te_ref| - |\Delta te| \quad dt_e = 0 \text{ if } |te| \geq te_ref$$

$$dt_e = -1 \text{ if } |te| \geq |te_ref| + \Delta te \quad t_e = 0 \text{ if } |te| \leq te_ref$$

Thus the digital output of flux comparator is 1,0 and that of torque comparator is 1,0,-1.

Depending on the flux and torque comparator

Sector 6: : $(\Psi_s ds < |\Psi_s qs) \ \& \ (\Psi_s ds \geq 0) \ \& \ (\Psi_s qs < 0)$

The voltage source inverter is modeled in form of S-function block, contained in voltage vector selection block. It is written in form of an M-file, which gives the output of the inverter in terms of three phase voltages as per the signals received from optimum pulse selector. In fact the inverter output voltages are reconstructed using dc link voltage in the table.

dΨ	D T	Sect.	Sect.	Sect.	Sect.	Sect.	Sect.
		1	2	3	4	5	6
1	1	V2	V3	V4	V5	V6	V1
	0	V7	V8	V7	V8	V7	V8
	-1	V6	V1	V2	V3	V4	V5
0	1	V3	V4	V5	V6	V1	V2
	0	V8	V7	U8	V7	V8	V7
	-1	V5	V6	V1	V2	V3	V4

V1, V2, V3, V4, V5 and V6 are active voltage switching vectors, V7 and V8 are zero voltage switching vectors. Optimum voltage vector selection is done using S-Function block contained in voltage vector selection block. This essentially constructs phase voltages using DC link voltage. The model also calculates sector (as shown in above table) in which the stator flux linkage space vector is lying as per following relations.

The model which is developed was simulated for an induction motor of rating 149.2 KW (machine parameters are given in appendix Simulation results in form of computer traces of electromagnetic torque developed, estimated torque , estimated and actual speed, stator flux and stator current have been depicted.

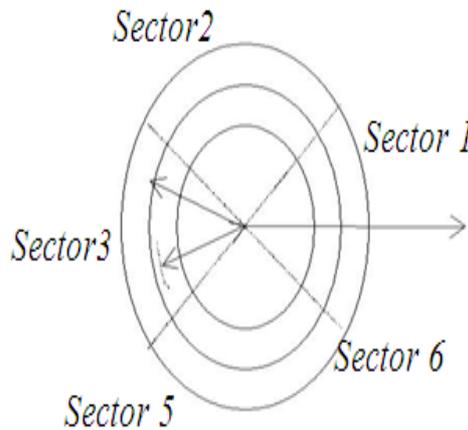


Fig. 7: Voltage vector selection block

SIMULATION RESULTS

The induction motor is fed by a voltage source inverter which is built using a Universal Bridge Block. The speed control loop uses a proportional-integral controller to produce the flux and torque references for the DTC block. The DTC block computes the motor torque and flux estimates and compares them to their respective reference. The comparators outputs are then used by an optimal switching table which generates the inverter switching pulses. The result of simulation is shown below:-When reference speed and load torque is

TEST1

Speed, at time(s):[0 1],amplitude is [300,0]

Torque, at time(s):[0 0.5 1],amplitude [0 692 -692]

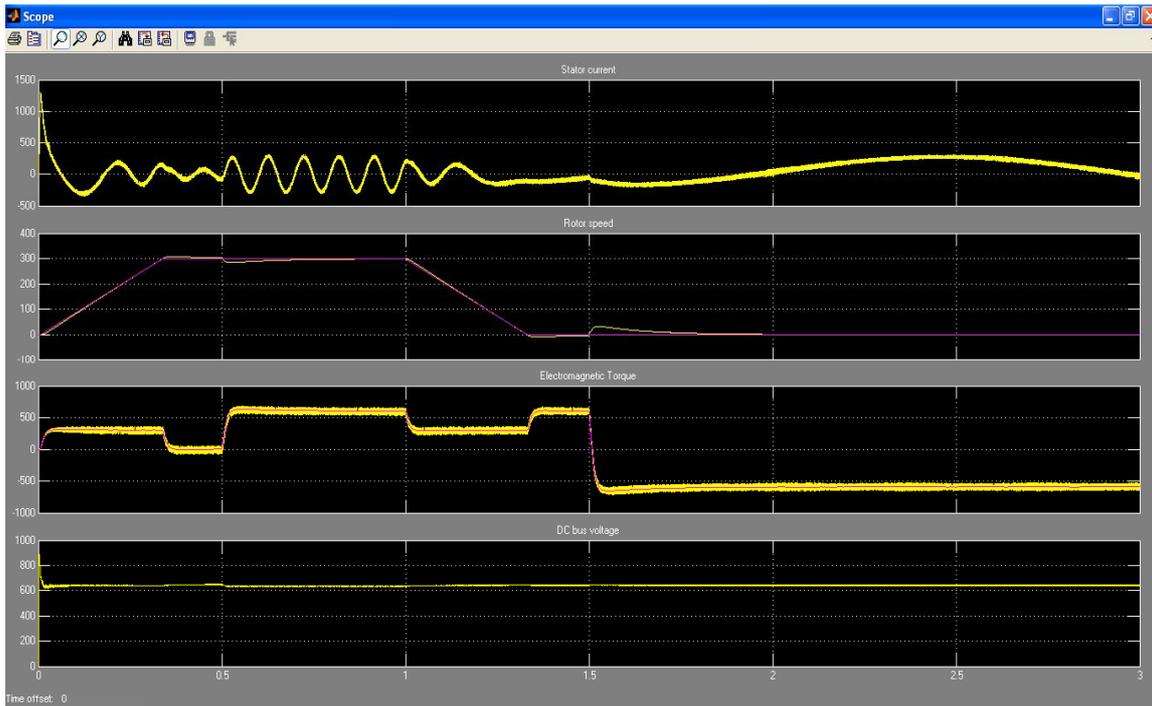


Fig. 8: simulation result

Start the simulation. You can observe the motor stator current, the rotor speed, the electromagnetic torque and the DC bus voltage on the scope. The speed set point and the torque set point are also shown.

At time $t = 0$ s, the speed set point is 300 rpm. Observe that the speed follows precisely the acceleration ramp. At time $t = 0.3$ s, speed is ramping to final value. The electromagnetic torque to decrease to 400 Nm and then stabilize at 600 Nm.

Condition:- The motor was started at no load with a set speed of 150 rad/sec and at $t = 0.5$ sec a load torque of 5 Nm.(65 % of rated torque) was applied . A PI controller is employed in the speed loop to make the steady state error in speed zero. The values of K_p and K_i are changed to obtain plots showing the dependence of the Simulation result of Stator current , Rotor speed , Electromagnetic torque and DC bus voltage performance. Estimated Flux remains the same in all the cases a shown in Fig. From then plots it is seen that estimated torque and electromagnetic torque are close to each other. Also actual and estimated speed plots are close to each other. By increasing the value of K_p , the pulsations in the electromagnetic torque increase.

Induction Motor

The three-phase squirrel-cage induction motor used for simulation and modeling has been detailed under this section. It has two inputs they are controlled phase voltage from DTC and load torque. The outputs of induction machine model are electromagnetic torque, speed, stator current, dc bus voltage. Parameters of Induction machine model are

$P_n = 200$ HP	$R_s = 14.85 \times 10^{-3} \Omega$
$U_n = 460$ V	$R_r = 9.295 \times 10^{-3} \Omega$
$p = 2$	$L_{ss} = 0.3027 \times 10^{-3}$ H
$J = 0.089$ kgm ²	$L_{rs} = 0.3027 \times 10^{-3}$ H
	$L_m = 10.46 \times 10^{-3}$ H

Mathematics designing of induction motor is done. The Induction motor model is given in Figure.

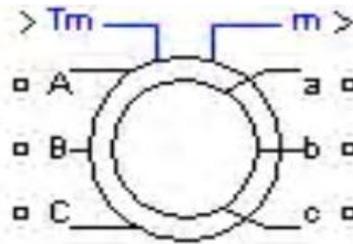


Fig. 9: Induction Motor

6.CONCLUSION

Direct torque control combines the benefit of direct flux and torque control into sensor less variable frequency drive that does not require a PWM modulator. Recent advances in digital signal processor and application specific integrated circuit and the theoretical concepts developed so far for direct self-control makes this possible. The objective of the present work was to make a model of direct torque control of three phase induction motor various speed control schemes were studied and extensive literature survey was carried out for understanding the direct torque control technique. MATLAB/SIMULINK was chosen as modeling and simulation tool because of its versatility. Model for direct torque controlled induction motor was developed using MATLAB/SIMULINK and performance of the system for different operating condition like starting, load changes, speed reversal, effect of changing the values of K_p and K_i on the performance characteristics, was studied. The model was validated by comparing the plots of various performance parameters. It was also observed that for motoring operation, the performance was best in terms of starting time, overshoot and undershoot.

REFERENCES

- [1] H. Kubota and K. Matsuse, "Speed sensorless field-oriented control of induction motor with rotor resistance adaptation," *IEEE Trans. Ind. Applicat.*, vol.30,pp.1219-1224,Sept./Oct.1994.
- [2] J.Maes and J. Melkebeek," Speed Sensorless direct torque control of induction Motor using and adaptive flux observer," *IEEE Trans. Ind. Applicat.*, vol. 36, pp. 33-37, May/June 2000.
- [3] Bimal K. Bose, "Modern Power Electronics and AC Drives" Ion Boldea, S.A. Nasar," *Electric Drives*".
- [4] I. Takahashi and T. Noguchi, "A new quickresponse and high efficiency control strategy of an induction machine," *IEEE Trans. Ind. Appl.*, vol. IA-22, pp.820-827, Sep./Oct. 1986.
- [5] U. Baader, M. Depenbrock, and G. Gierse, "Direct self control(DSC) of inverter-fedinduction machine—A basis for speed control without speed measurement," *IEEE Trans. Ind. Appl.*, vol. 28, pp.581-588, Jun. 1992.